Simulation of electromagnetics absorption in human head for mobile telephone at 900 MHz and 1800 MHz

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Department of Optometry, Faculty of Medical Technology, Surman, Zawia University **Abstract**

In this paper, Specific Absorption Rates (SAR) inside the human head and hand have been analyzed for a handheld mobile phone with Planar Inverted-F Antenna (PIFA) is used to radiate dual-band frequency of GSM 900/1800 MHz in the vicinity of the SAM head phantom. Simulations are performedusing Three-dimensional Finite Integral technique (FIT) method via CST-Microwave studio has been used to simulate SAR induced in the head and mobile. Maximum peak 1-g and 10-g SARs of 0.5073 W/kg and 0.5479 W/kg are found at antenna resonance frequency of 900 MHz for 0.125 W applied input power respectively, while maximum peak 1-g and 10-g SARs of 0.762 W/kg and 0.9041W/kg at 1800 MHz for 0.250applied input power respectively. All simulations are below the limits set by ANSI/IEEE and FCC.

1.Introduction

In recent years, interest has been paid to the potential health hazards resulting exposure electromagnetic radiation (EM) and to especially the head region. Absorption of Radio Frequency (RF) fields emitted from mobile phone the may change the proliferation rate of cells, enzyme activity and affect the genes in the DNA of cells and may form tumor in living tissues [1]. It has also been reported that the opening of the blood brain barrier due to low level EM radiation emitted from a mobile phone causes to release the dangerous chemicals into the brain, leak hemoglobin and building up of which can cause heart diseases and kidney stones [2].the specific absorption rate (SAR) is a physical quantity that used to evaluate the power absorbed by biological

tissue. SAR is used to quantify biological adverse effects and formulating safety guidelines orstandards on exposure to RF fields [3-4]. The guideline that provides SAR exposure limit of 1.6 W/kg for any 1 g oftissue was approved by the IEEE in 1991 and was subsequently adopted by the American National Standards Institute (ANSI) in 1992 as a replacement for the previous (ANSI C95.1-1982 guideline). In April 1993, the FCC proposed using the ANSI/IEEE C95.1-1992 for evaluating RF environmental fields created by transmitters it licenses and authorizes [6,7]. IEEE C95.1-2005 the newly approvedstandard represents a complete revision of and replaces IEEE Standard C95.1-1991[6]. The SAR limit specified in IEEE C95.1: 2005 has been updated to 2 W/kg over any 10-g of tissue. IEEE C95.3 is а recommended practice for measurements and computations of radio frequency electromagnetic fields with respect to human exposure to Such Fields, 100 kHz to 300 GHz. Direct measurement of SAR is very difficult inside a living human head or body parts using the experimental technique. Therefore, simulations using numerical techniques are used to calculate EM field

2.Physical Model.

In this study, a mobile phone with a Planar Inverted-F Antenna (PIFA) was used.The reference power of the phone was 0.125 W and 0.250 W, defined according to the Standard IEEE C.95.3.The mobile phone is located at the right side of a human head with a certain position.A near-field radiation source for human head model is considered. components and SAR inside human head or body parts [8].The work described in this paper is substantially extended from our previous work [9]. In this studya 3D handset together with the SAM phantom model including handwas used to simulate the SAR distribution over the human head. A handset with a Planar Inverted-F Antenna (PIFA) was used. The 900 MHz and 1800 MHz frequencies were chosen for the simulations in this study.Finite integration in timedomain (FITD) method is used [10-11,13].

Fig. (1) shows the simulation model which includes the handset with PIFA type of the SAM phantom head antenna and CST Microwave provided by Studio Anthropomorphic [12].Standard Model (SAM) head is a homogeneous model of the human head composed of two parts: fluid and shell as shown in Fig. 2.



The simulated handset model is composed of the circuit board, LCD display, keypad, battery and housing. This handset is composed of A dual-band Planar Inverted-F Antenna (PIFA) operates at 900 MHZ and 1800 MHZ is used as the radiating antenna, which is shown in Fig. (3).The electrical properties of materials used for simulation are listed in Table (1).

	Phone Materials	ε _r	σ (S/m)
	Circuit Board	4.4	0.05
	Housing Plastic	2.5	0.005
	LCD Display	3.0	0.02
	Rubber	2.5	0.005
	SAM Phantom Head	ε _r	σ (S/m)
	Shell	3.7	0.0016
	Liquid @ 1.8 GHz	40	1.42
Fig. (3) The provided dual-band Planar	Table (1) Electrical properties of materials		
Inverted-F Antenna	used for simulation		

3. Numerical Method

The Finite Integral Time-Domain technique (FITD) proposed by Weiland in 1976 [14], CST Studio software package was used as the main simulation tool.Maxwell's equations and the related material equations are transformed from the continuous to the discrete spaceby allocating electric voltages on the edges of a grid and magnetic voltages on the edges of a dual grid[15-16]. Fig. (4) depicts the allocation of the electric grid voltages e and magnetic facet fluxes b on the primary grid G. In addition, the dielectric facet fluxes d as well as themagnetic grid voltages h are defined on the dual grid \tilde{G} (indicated by the tilde):



 $\begin{array}{ll} e_i: electric \ voltage & h_i: magnetic \ voltage \\ b_i: magnetic \ flux & d_i: electric \ flux \\ Fig.(4). \ Principle \ of \ FIT \ calculation \ [17] \end{array}$

The spatial discretization of Maxwell's equations is finally executed on these two orthogonal grid systems. The electric grid voltages (e), magnetic facet fluxes (b) are allocated on the primary grid (G). In addition, the dielectric facet fluxes (d) as well as the magnetic grid voltages (h) are defined on the dual grid (shown by the tilde). Applying this scheme to Ampere's

law on the dual grid requires identification of a corresponding dual discrete curl operator (C tilde). Similarly, the discretization of the remaining divergence equations introduces discrete divergence operators (S and S tilde), belonging to the primary and dual grids, respectively. As previously indicated (Fig. 5), these discrete matrix operators consist of elements '0', '1' and '-1', representing merely topological information. Finally, the complete discretized set of Maxwell's Grid Equations (MGEs) is set up (Fig. 6)



Maxwell's equation	Maxwell's grid equation (MGEs)		
$\oint_{C} \vec{H} \cdot d\vec{l} = \int_{S} \left(\vec{J} + \frac{\partial \vec{D}}{\partial t} \right) d\vec{S}$	$\widetilde{\boldsymbol{C}}\boldsymbol{h} = rac{d}{dt}\boldsymbol{d} + \boldsymbol{j}$		
$\oint_{c} \vec{E} \cdot d\vec{l} = -\int_{s} \frac{\partial \vec{B}}{\partial t} d\vec{S}$	$oldsymbol{C}oldsymbol{e}=-rac{d}{dt}oldsymbol{b}$		
$\oint_{C} \vec{D}.d\vec{S} = -\int_{S} \rho dV$	$\widetilde{S}d = q$		
$\oint_C \vec{B} \cdot d\vec{S} = 0$	Sb = 0		

Fig.(6) Maxwell's equations and algebraic matrixequations.

The main benefit of FIT is the possibility to have two different materials within one grid cell, whereas in FDTD only one material is allowed within one grid cell. Due to this advantage, the mesh can be significantly sparser, and hence, less memory is required in FIT simulations, especially in the objects with complex geometry [11]. Fig.(7)shows the mesh

view for two cut planes of the complete model indicating the area with denser meshing along the inhomogeneous boundaries.



Fig. (7) Mesh view for two cut phones of the complete model showing the non-uniform meshing scheme adopted for simulation.

SAM phantom head was then included for SAR calculation using the standard definition as

$$SAR = \frac{\sigma}{2\rho} E^2$$

where E is the induced electric field (V/m); ρ is the density of the tissue (kg/m³) and σ is the conductivity of the tissue (S/m). The corresponding SAR values averaged over 1 g and 10 g of tissue in the head were denoted as SAR_{1 g} and SAR_{10g}, respectively. These values were used as a benchmark to evaluate the effectiveness in peak SAR reduction.

4. Simulation results

4.1. PIFA antenna simulation

The numerical calculation of the phone with PIFA, were done as follows: f = 900 and 1800 MHz, P = 0.125 W, and 0.250 W respectively, and $Z = 50 \Omega$. The distance from the head (ear) to the edge of the mobile phone was 1 cm. The PIFA antenna is simulated CST SOFTWARE. The parameters evaluated were gain, beamwidth and return loss. Figure 5 shows the simulated S11 of the complete model including handset and the SAM phantom head. Figures (8-10) present simulation of 3D farfield radiation patterns at 900 and 1800 MHz. Fig. (8-9) simulatePIFA mobile antenna parameters S_{11} namely: return loss, radiation efficiency, total efficiency and directivity, the results obtained with the presence of the human head and at a frequency 900 and 1800 MHz. The return loss for 900 MHz shows a drops around -4 dB, while the 1800 MHz also shifts to the right but with less loss at -5.5 dB



Fig.(8). Simulated *S*11 of the complete model including handset and SAM phantom head showing proper operation of the antenna.

The 3D farfield radiation patterns at 900 and 1800 MHz together with are included in Figs. 9-10



Figure 9. Directivity for 900 and 19800 MHz resonance frequencies (3D directivity pattern has been magnified for better presentation).

Fig. (10) presents the polar plots at two corresponding cuts for the two frequencies.



Fig.10. Simulated 2D radiation patterns of the PIFA antenna in the presence of head and hand on the XY, YZ and XZ planes at 900MHZ (absolute gain).

4.2. SAR simulation

The electromagnetic field emitted by PIFA mobile phone antenna is simulated at two GSM frequencies (900 MHz and 1800 MHz) to see the 3D dimension effect of SAR penetrating the head phantom averaged over a mass of 10 gram and 1 gram cubic. Table (2)shows the results of

Operati	ing Frequency	900 MHz	1800 MHz
SAR 10 g	ear view (xy plane)	0.340 0.340 0.240 0.340 0.2400 0.2400 0.2400 0.240000000000	0.53 0.54 0.55 0.55
	cutplane view (xz plane)		0.27 b 0.27 b 0.26 b 0.46 b 0.
	Max SAR :	0.5479 (W/Kg)	0.9041 (W/Kg)
SAR 1 g	ear view (xy plane)	0.777 0 0.777 0 000 0 0000 0 0000000000	0.42 € 10 0.42 € 10 0.42 € 10 0.42 € 10 0.45 € 100
	cutplane view (xz plane)	0.389 0.307 × 0.007 × 0.000 0.389 14 0.0000 0.389 14 0.0000000000000000000000000000000000	1.492 1.493
	Max SAR :	0.5073 (W/Kg)	0.762 (W/Kg)

maximum SAR value in *XY* -plane (ear view) and cutplane-view of *XZ*-plane to evaluate at both axes the absorption direction into the head.

Simulation results in tables (2)show that the 1-g and 10-g SAR values (cutplane view xz plane) arelarger as frequency increases while penetrationdepth decreases with increasing frequency. Also, large SARvalues are observed on the head-air interface at highfrequencies (900MHz and 1800 MHz) due to theimmediate change of the refraction index from thevalue 1 in air to a much larger value in brain matter. Tables (2) also showed variations of maximum 1-g and 10-g SARs with varies distances along Y-axis in the mid-coronal plane at 900 MHz and 1800 MHz.

IV. CONCLUSION

In this work SAR distributions and peak SAR averaged over 1-g and 10-g mass of head phantom exposed to a mobile phone designed for two different frequencies (900-1800 MHZ) have been studied using FIT method. Calculation of SAR has been performed using commercially available software CST MWS. At 900 MHz, variation of peak 1-g and 10-g SARs with distance shows that both maximum 1-g and 10-g SAR value reach to maxima near the position of the mobile phone antenna and decreases gradually with increase of the distance from the mobile phone antenna. Results obtained by the simulation show that maximum peak of 10-g SAR obtained in head surface phantom at 900-1800 MHZ are 0.5479 (W/Kg)and 0.9041 (W/Kg)respectively whereas 0.5073 (W/Kg) and 0.762 (W/Kg) peak for 1-g. Variations of peak 1-g and 10-g SARs with varies distances along Y-axis in the mid-coronal plane at 900-1800 MHZ obtained using CST-Microwave studio are observed. Simulated peak 1-g and 10-g SARs for human head with hand held mobile is compared with measured SARs available in the literature and it is observed that obtained simulated and measured SAR values are close each other and to

lower than the corresponding measured values within the ANSI/IEEE and FCC safetylimits

V. References

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